Thermal Modelling & Temperature Prediction of a Power Module Mounted on an IMS PCB

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My Project

- •Connected to a current 'Multi Physics Sensor Fusion' research project between several universities.
- •Developed a thermal model for a simple power module mounted on an insulated metal substrate (IMS) PCB.
- •Model can predict temperatures of all devices for any given power dissipation in any device combination.

- •Rise in development of electric vehicles.
- •Increasing demand for power electronics with greater power capabilities with smaller physical footprints.
- •Complicates thermal management:
	- Devices close together thermal cross coupling.
	- High operating temperatures.
	- Rapid thermal cycling.
	- Mechanical stress.
	- Cracked solder joints.
	- Bond wire lift off.
- •Thermal model can be used in conjunction with degradation model.

Project Components

Theory: Temperature Measurement

- •External: Thermocouples
	- Fast response time.
	- Inexpensive and widely available.
	- Require complex electronics.
	- Additional thermal impedance between case and semiconductor die – time delay.

- •Internal: Electro-thermal relationships
	- Takes advantage of intrinsic features in devices structure.
	- MOSFET: Anti-parallel diode / body diode.
	- If current is constant, temperature is function of diode voltage.

$$
V_d = \frac{kT}{e} \ln\left(\frac{I}{I_s} - 1\right)
$$

- Direct die temperature measurement.
- Linear relationship over range of interest (30°C 120°C)

$$
V_d = m \cdot T_j + V_0
$$

Theory: RC Thermal Model

- •Conventionally, thermal resistance used to calculate heatsink value.
- •Does not describe transient behaviour capacitors added to produce complex thermal impedance.
- •Modelled in RC network form computationally lightweight/LTSpice

•Foster Network:

- R and C values found from system step response.
- Time constants of each section found from fitting step response to multi-term exponential equation.

Theory: RC Thermal Model

- •Step response found by applying step change in power dissipation and observing the temperature rise.
- •Converted into a transient thermal impedance curve.

$$
Z_{TH}(t) = \frac{T_j(t) - T_a(t)}{P}
$$

- •Fitted to multi-term exponential curve.
- •RC values extracted.

$$
Z_{TH}(t) = \sum_{n=1}^{N} R_n \left[(1 - \exp\left(-\frac{t}{R_n C_n}\right) \right]
$$

•Transfer function of network derived.

$$
Z_{TH}(s) = \frac{R_1}{1 + sR_1C_1} + \frac{R_2}{1 + sR_2C_2}
$$

•Total junction temperature elevation due to both self heating and thermal cross coupling.

•Linear superposition of contributions from heat sources.

Self heating Cross coupling
\n
$$
\begin{bmatrix}\nT_{j1} \\
T_{j2} \\
T_{j3} \\
T_{j4}\n\end{bmatrix} = \begin{bmatrix}\nT_a \\
T_a \\
T_a \\
T_a\n\end{bmatrix} + \begin{pmatrix}\nZ_{11} & Z_{12} & Z_{13} & Z_{14} \\
Z_{21} & Z_{22} & Z_{23} & Z_{24} \\
Z_{31} & Z_{32} & Z_{33} & Z_{34} \\
Z_{41} & Z_{42} & Z_{43} & Z_{44}\n\end{bmatrix} \times \begin{bmatrix}\nP_1 \\
P_2 \\
P_3 \\
P_4\n\end{bmatrix}
$$

Method: Circuit

- •Needed a circuit capable of precisely controlling the power dissipation in a device.
- •Vin maintained across Rbias, 2Vin maintained across test MOSFET.
- •Hence Vin also maintained across test MOSFET.
- •Current set by Rbias.
- •Dissipation in resistor equal to dissipation in test MOSFET.

- •Needed circuit to measure body diode voltage.
- •Gate of MOSFET grounded so current only flows through body diode.
- •Constant current driver forward biases body diode with ~5mA – Develops voltage between source and drain.
- •Differential amplifier outputs diode voltage into ADC.

Method: Combined Circuit

•Voltage measurement circuit operates on an isolated supply since it has to reverse bias the MOSFET.

Method: Simultaneous Operation

•Both circuits cannot operate simultaneously.

- •PWM style switching scheme used to rapidly switch between circuits.
- •Circuit designed onto Arduino Uno 'shield'.
- •Arduino communicates to data acquisition program.

Method: Data Acquisition Program

Method: Diode Voltage Calibration

- •IMS PCB placed onto temperature controlled hotplate.
- •Diode voltage measured from 30°C 100°C.

- •Thermocouples did not reach temperature of semiconductor die.
- •MOSFET suspended in thermally insulated container.
- •Even heating by surrounding ambient air.

Results: Diode Voltage Calibration

- •Very low error at room temperature (<1°C).
- •Approximately 10°C difference at equilibrium with 10W dissipation
- •Caused by thermal impedance between semiconductor die and case.
- •1°C/W would cause a temperature drop of 10°C due to a dissipation of 10W.
- •Highlights the uncertainty caused by external temperature measurement techniques.

Method: Step Response

- •Circuit only capable of measuring diode voltage of one device at a time.
- •Thermocouples were used for the modelling process.
- •Battery powered laptop used to achieve ground isolation between power controller and voltage measurement circuits.
- •Performed 10W step power dissipation in each device in turn.
- •Voltage at MOSFET terminals and current measured to calculate power.
- •Measured and recorded temperature elevation of all devices.

Results: Step Response

- •Temperature curve converted into thermal impedance curve.
- •Error possibly caused by to additional thermal impedances caused by mounting bolts and pin headers.
- •Curve fit obtained for each curve.
- •Through experimentation, four exponential terms were suitable to produce a precise curve fit.

- •First term made negative to account for non-instantaneous temperature rise.
- •Improved resemblance to measured curve shape.

Results: Step Response

Model equation for each thermal path

$$
Z_{TH}(t) = -R_1 \left[1 - exp\left(-\frac{t}{R_1 C_1}\right) \right] + R_2 \left[1 - exp\left(-\frac{t}{R_2 C_2}\right) \right] + R_3 \left[1 - exp\left(-\frac{t}{R_3 C_3}\right) \right] + R_4 \left[1 - exp\left(-\frac{t}{R_4 C_4}\right) \right]
$$

RC vales for each thermal path

Transfer function of each thermal impedance entered into MATLAB using tf() command.

%Thermal impedances due to dissipation in device 1

Z11 = tf(-Z11 R1/(1+s*Z11 R1*Z11 C1) + Z11 R2/(1+s*Z11 R2*Z11 C2) + Z11 R3/(1+s*Z11 R3*Z11 C3) + Z11 R4/(1+s*Z11 R4*Z11 C4)); Z21 = tf(-Z21_R1/(1+s*Z21_R1*Z21_C1) + Z21_R2/(1+s*Z21_R2*Z21_C2) + Z21_R3/(1+s*Z21_R3*Z21_C3) + Z21_R4/(1+s*Z21_R4*Z21_C4)); Z31 = tf(-Z31 R1/(1+s*Z31 R1*Z31 C1) + Z31 R2/(1+s*Z31 R2*Z31 C2) + Z31 R3/(1+s*Z31 R3*Z31 C3) + Z31 R4/(1+s*Z31 R4*Z31 C4)); Z41 = tf(-Z41_R1/(1+s*Z41_R1*Z41_C1) + Z41_R2/(1+s*Z41_R2*Z41_C2) + Z41_R3/(1+s*Z41_R3*Z41_C3) + Z41_R4/(1+s*Z41_R4*Z41_C4));

Results: Modelling

- •Square wave with variable frequency and duty cycle used to test model.
- •Power dissipation defined for each device.
- •lsim(Z, P, t) command used to extract temperature in time domain from transfer function.

- •Large error between model and measurements.
- •Caused by non-constant ambient air temperature.
- •Experiment repeated with ambient air thermocouple.
- •Ambient air model added to current model.

- Tested again at 100mHz with a 50% duty cycle.
- Improved accuracy of model compared to measured results.
- Highlights the effects of thermal cross coupling.

- •5W dissipation in device 1 & 4.
- •Excellent agreement between model and measured data.

Conclusion

•Internal temperature measurement technique has been presented.

- •Methodology to thermally characterise a board and produce a simple thermal model has been described.
- •Model has been tested and compared to measured data good correlation.

Thank you for listening…